

# Application Note

## Selecting Input and Output Capacitors with Dialog's Integrated Power Switches

### AN-CM-251

#### **Abstract**

*This application note is devoted to the topic of selecting proper capacitors to use with Dialog's Integrated Power Switches. Important considerations include desired electrical performance, system transient requirements, load parameters, and voltage deviation specifications.*

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Selecting Input and Output Capacitors with Dialog's Integrated Power Switches

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## Selecting Input and Output Capacitors with Dialog's Integrated Power Switches

### 1 Terms and Definitions

ACL	Active Current Limit
ESL	Equivalent Series Inductance
ESR	Equivalent Series Resistance
IPS	Integrated Power Switch
SCL	Short-circuit Current Limit

### 2 References

- [1] SLG59M1714V, Datasheet, Dialog Semiconductor
- [2] AN-1207, Inductive Load Power Switch Considerations, Dialog Semiconductor
- [3] AN-1068, GFET3 and HFET1 Integrated Power Switch Basics, Dialog Semiconductor.
- [4] AN-CM-246, Applying Dialog Integrated Power Switches in Super Capacitor Applications, Dialog Semiconductor.

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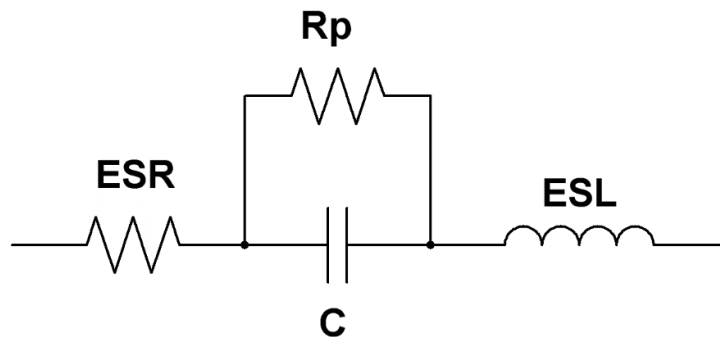
### 3 Introduction

Capacitors are used in almost all electronic products in a variety of ways. Capacitors provide a number of essential functions in circuit design, such as providing flexible filter options, noise reduction, power storage, and sensing capabilities. To effectively use capacitors in power switch designs, some key details need to be taken into account. Desired electrical performance, system transient requirements, load parameters, and voltage deviation are very important to consider when selecting proper input and output capacitors in a given application.

Dialog Semiconductor offers a wide range of Integrated Power Switches (IPS) for a variety of applications. For more information, please visit [www.dialog-semiconductor.com/power-switching](http://www.dialog-semiconductor.com/power-switching).

### 4 Capacitor Parasitic Effects

While an ideal capacitor is capable of transferring all its stored energy to a load instantaneously, a real capacitor has parasitic components that prevent this behavior. An equivalent circuit model of a real capacitor is illustrated in Figure 1. As shown, the capacitor equivalent circuit comprises four elements: capacitance, an equivalent series inductance (ESL), a high-resistance DC path ( $R_p$ ) in parallel with the capacitance, and an equivalent series resistance (ESR).



**Figure 1. Capacitor Equivalent Circuit**

The electrodes and the leads of a capacitor contribute the resistive and inductive components while its dielectric material and its construction contribute to the insulation resistance.

High ESR degrades performance due to  $I^2R$  losses, noise, and larger voltage drop. On the other hand, ESL causes a magnetic field to buildup in capacitor. The buildup of magnetic field interferes with how current peaks and recovers. Both ESR and ESL depend on the type of capacitor and its construction.

Wet aluminum electrolytic capacitors are used primarily for bulk decoupling applications. However, their relatively high ESR and ESL slow response times and reduce performance.

Aluminum polymer capacitors have better performance characteristics, and they are increasingly replacing wet aluminum capacitors in bulk decoupling applications. Aluminum polymer capacitors exhibit much lower parasitic ESR and ESL.

Tantalum capacitors are a subclass of electrolytic capacitors. They are made of tantalum metal which acts as an anode, covered by a layer of oxide which acts as the dielectric, then surrounded by a conductive cathode. Tantalum capacitors have an equivalent series resistance (ESR) ten times smaller than the ESR of aluminum electrolytic capacitors, which allows for larger currents to pass through with less heat generated and, in addition, smaller, parasitic IR voltage drops. Tantalum

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capacitors are very stable over time and their capacitance doesn't change with age significantly, especially when compared to aluminum electrolytic capacitors.

Ceramic capacitors are most commonly used in electronic circuits for decoupling applications. They have relatively low equivalent series resistance, but their ESL is greatly determined by the distance between terminations (its construction).

If PCB space is not an issue, capacitors can be connected in parallel to reduce both ESR and ESL while beneficially increasing the effective capacitance.

## 5 How to Select Capacitors

In a general sense, Dialog's IPSs don't require any input or output capacitors. The use of input and output capacitors is determined by the usage scenario (application) Typically, to turn on a Dialog IPS,  $V_{DD}$ ,  $V_{IN}$ , and ON signals are applied. Every Dialog IPS datasheet contains an information on the proper sequencing of these three signals. In short  $V_{DD}$  should be applied first, then  $V_{IN}$ , and finally the ON signal can be toggled low-to-high (for asserted-HIGH ON signals or high-to-low for asserted-LOW ON signals) to close the switch. Also, it is recommended that  $V_{DD}$  and  $V_{IN}$  rise times should be longer than 2 ms. As an example a typical power up operation of SLG59M1714V [1] is illustrated in Figure 2.

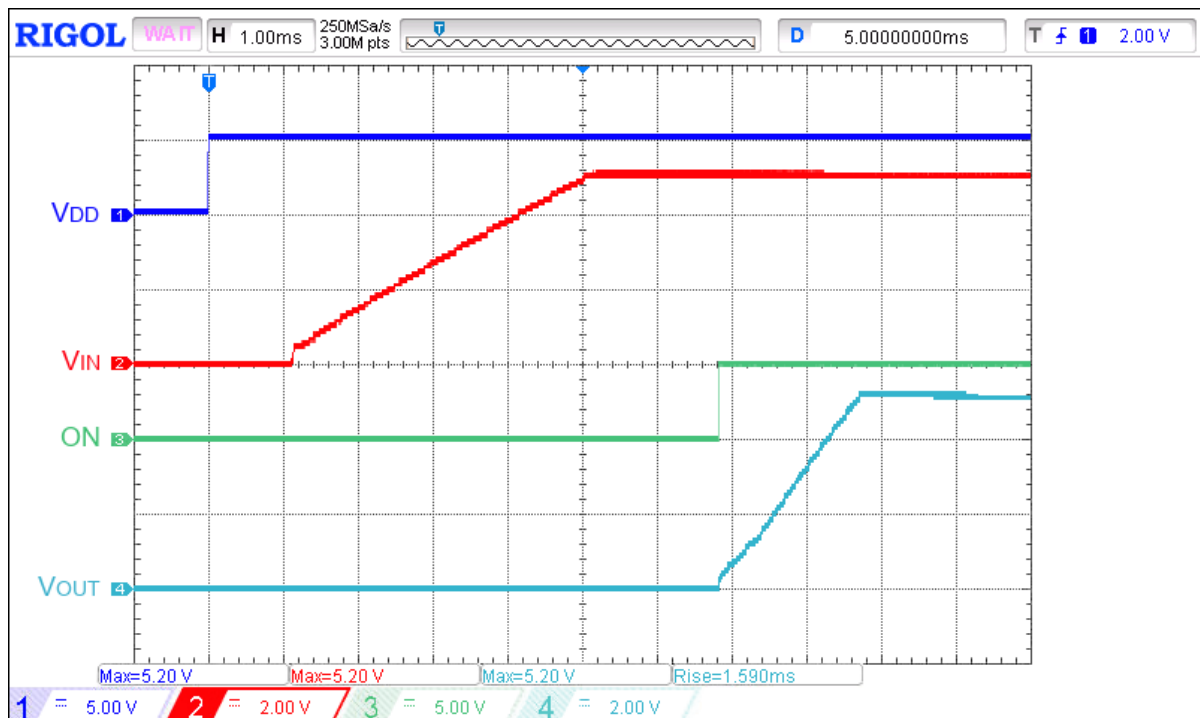


Figure 2. A Typical Power Up Behavior of a SLG59M1714V.

However, there are some applications when  $V_{DD}$  and  $V_{IN}$  have fast rise times (Figure 3), or when  $V_{DD}$  and  $V_{IN}$  are applied simultaneously (Figure 4), or when  $V_{IN}$  is applied before  $V_{DD}$  (Figure 5). In each of these cases, a voltage glitch may appear at the output even when ON = GND (for asserted-HIGH ON signals).

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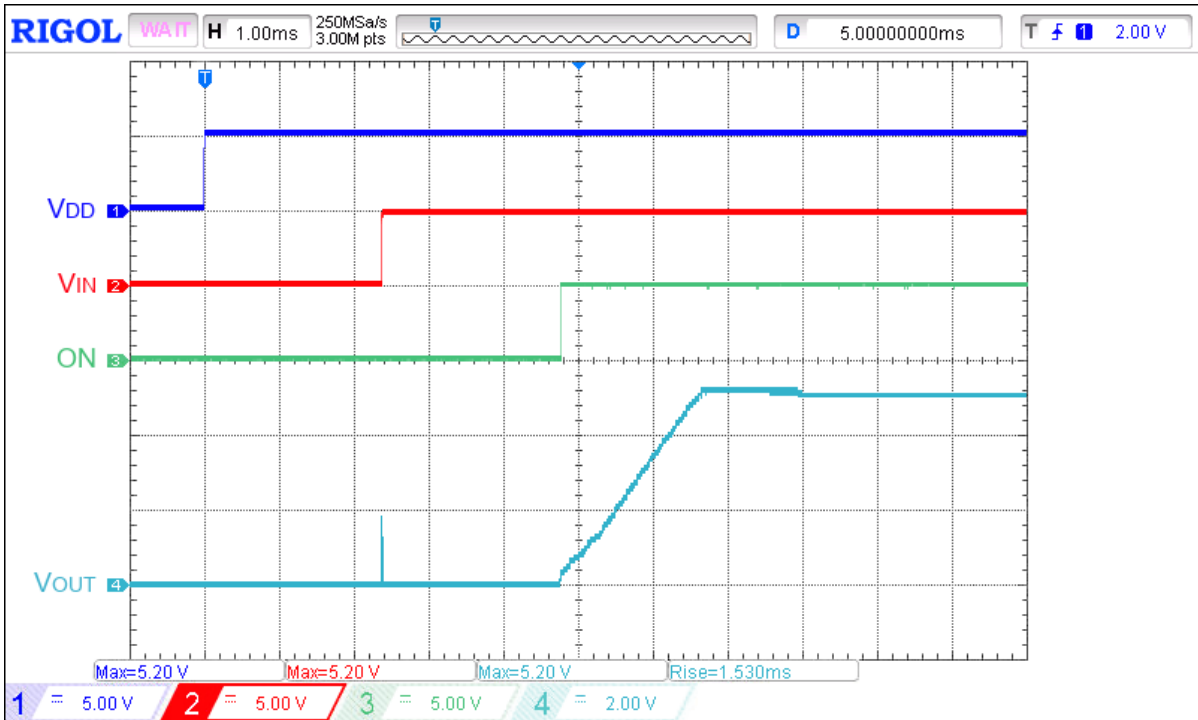


Figure 3. Powering zup an IPS with Fast  $V_{DD}$  and  $V_{IN}$  rise Times.

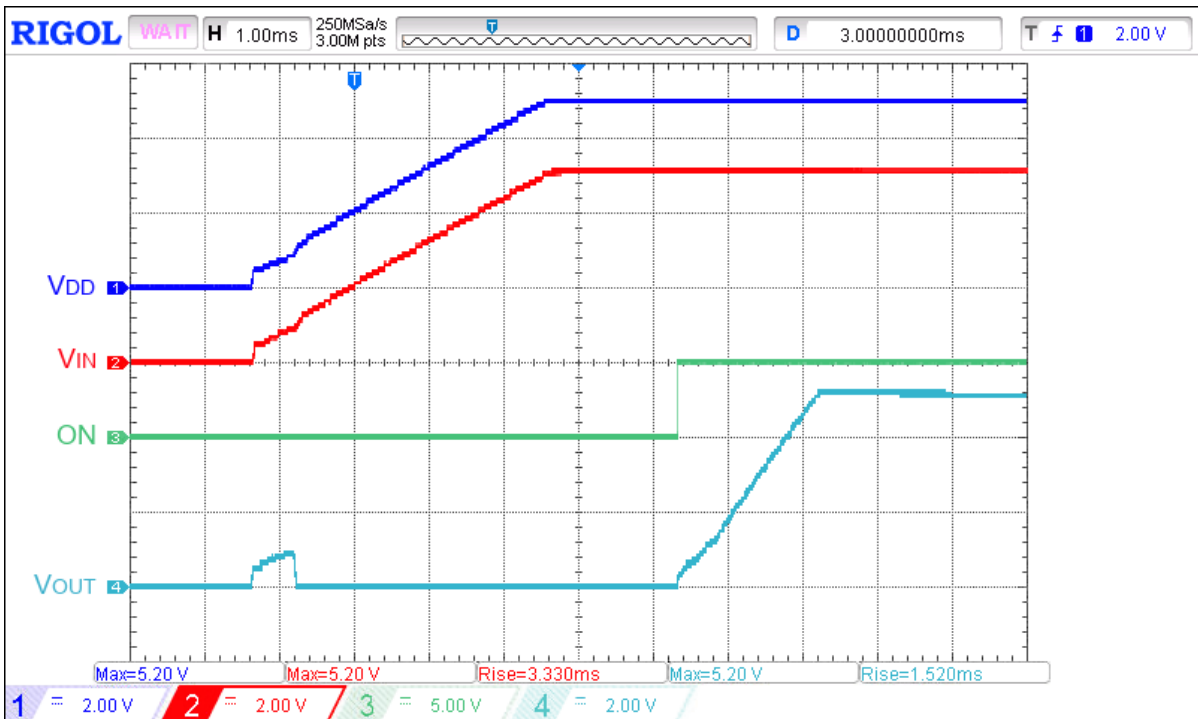


Figure 4. Powering Up an IPS when  $V_{DD}$  and  $V_{IN}$  are Applied Simultaneously.

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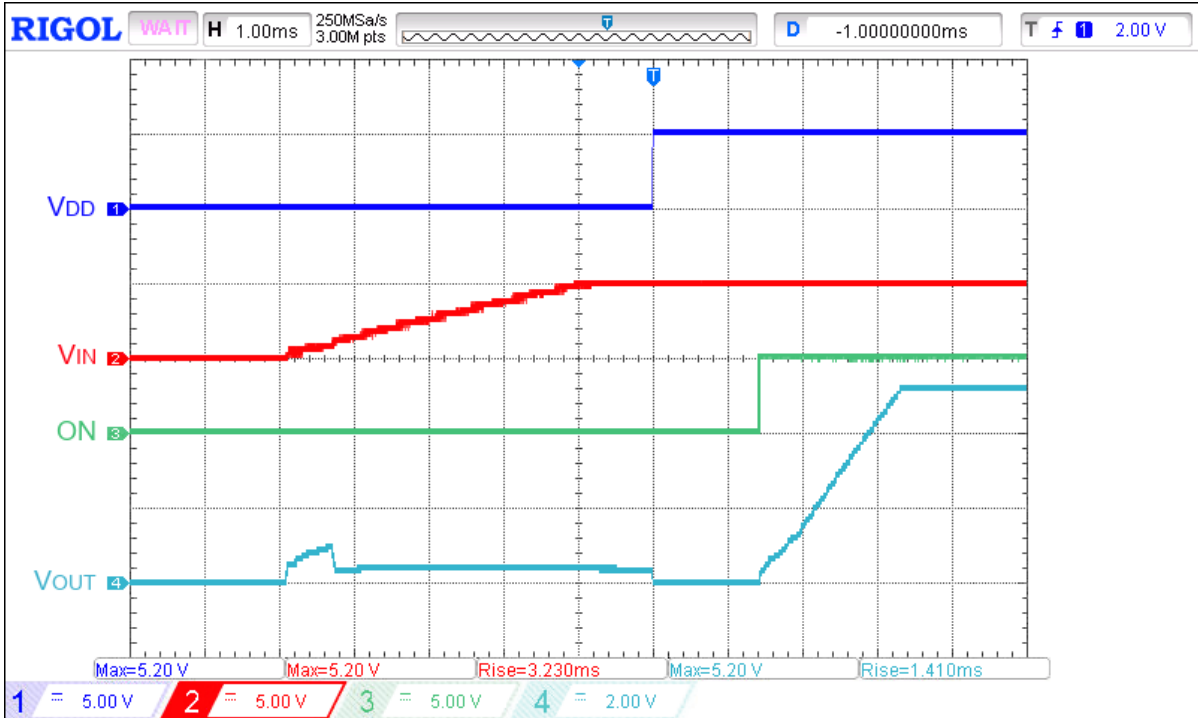


Figure 5. Powering Up an IPS when  $V_{IN}$  is Applied Before  $V_{DD}$ .

To avoid such glitches, an output capacitor ( $C_{LOAD}$ ) or a resistor ( $R_{LOAD}$ ) should be added at the downstream side of the IPS. In this particular case, a  $1\mu F$  ceramic capacitor at  $V_{OUT}$  helps to get rid of that glitch (Please see Figure 6, Figure 7, and Figure 8, inclusive).

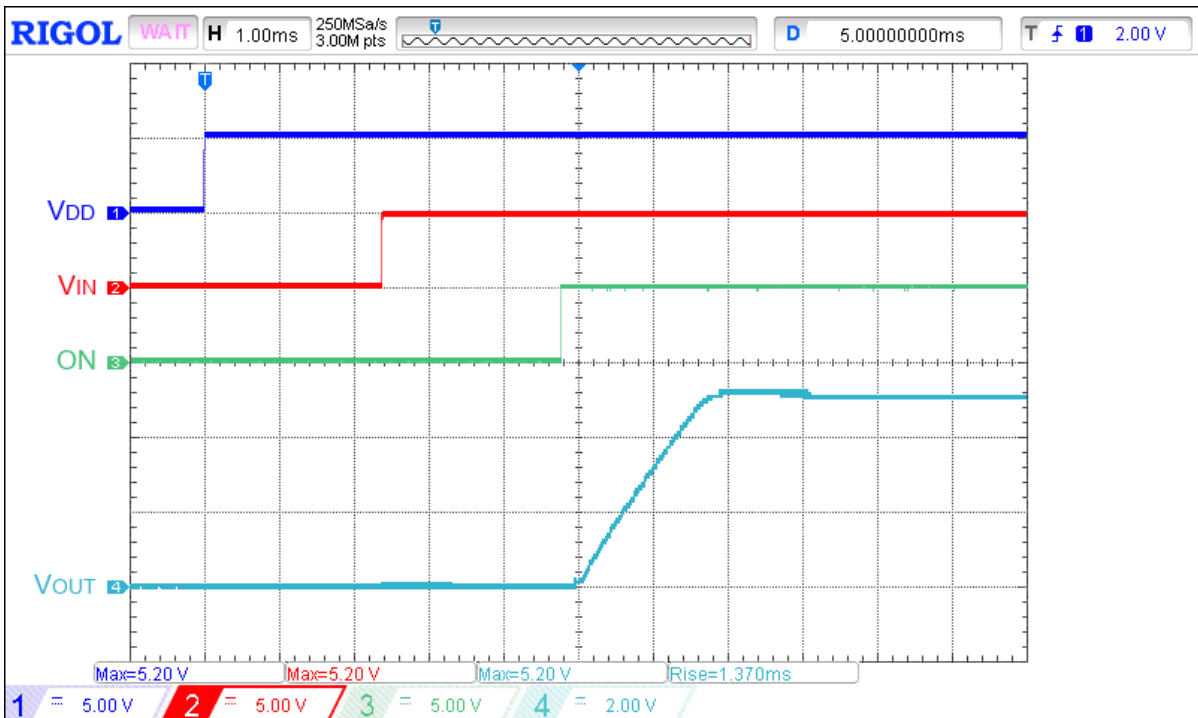


Figure 6. An IPS Powering Up a  $1\mu F$  Load Capacitor with Fast  $V_{DD}$  and  $V_{IN}$  rise Times.

## Selecting Input and Output Capacitors with Dialog's Integrated Power Switches

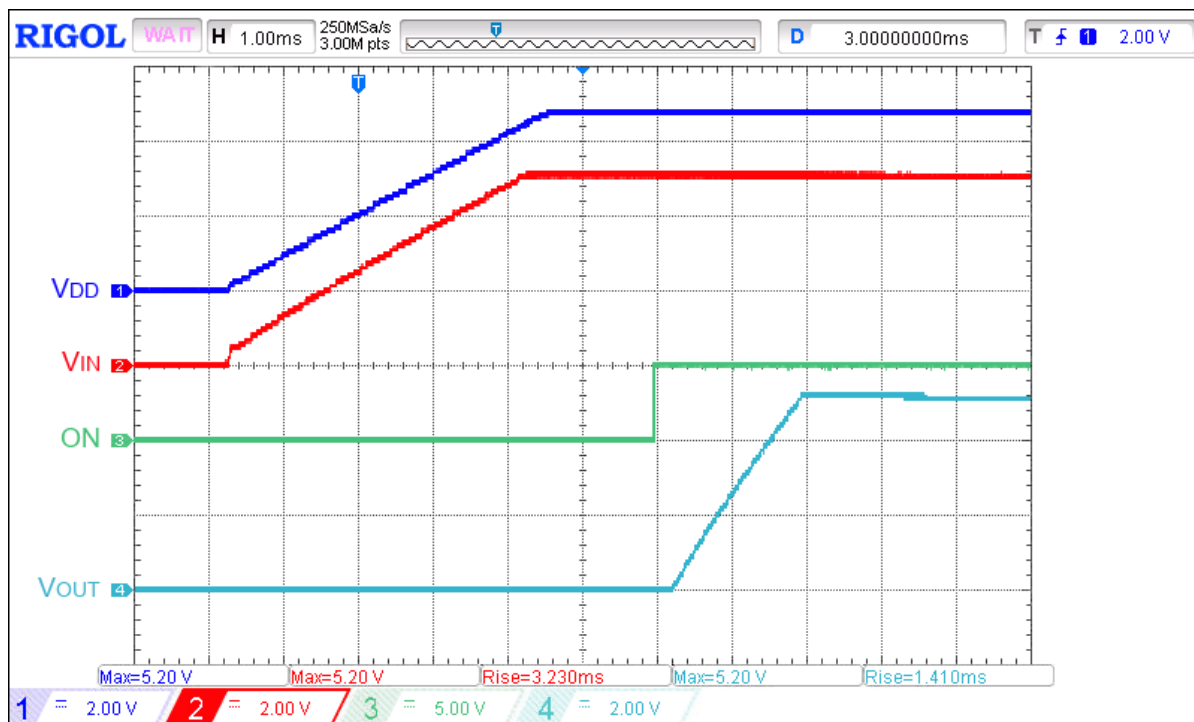


Figure 7. An IPS Powering Up a 1  $\mu$ F Load Capacitor when  $V_{DD}$  and  $V_{IN}$  are Applied Simultaneously.

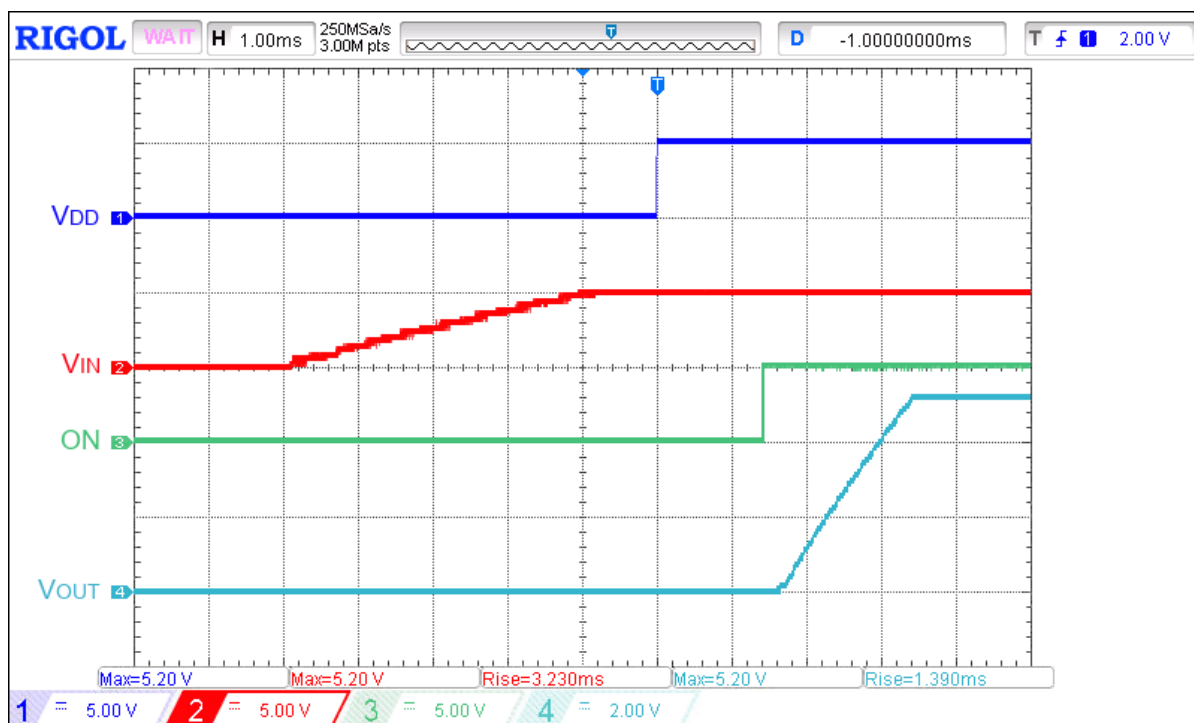
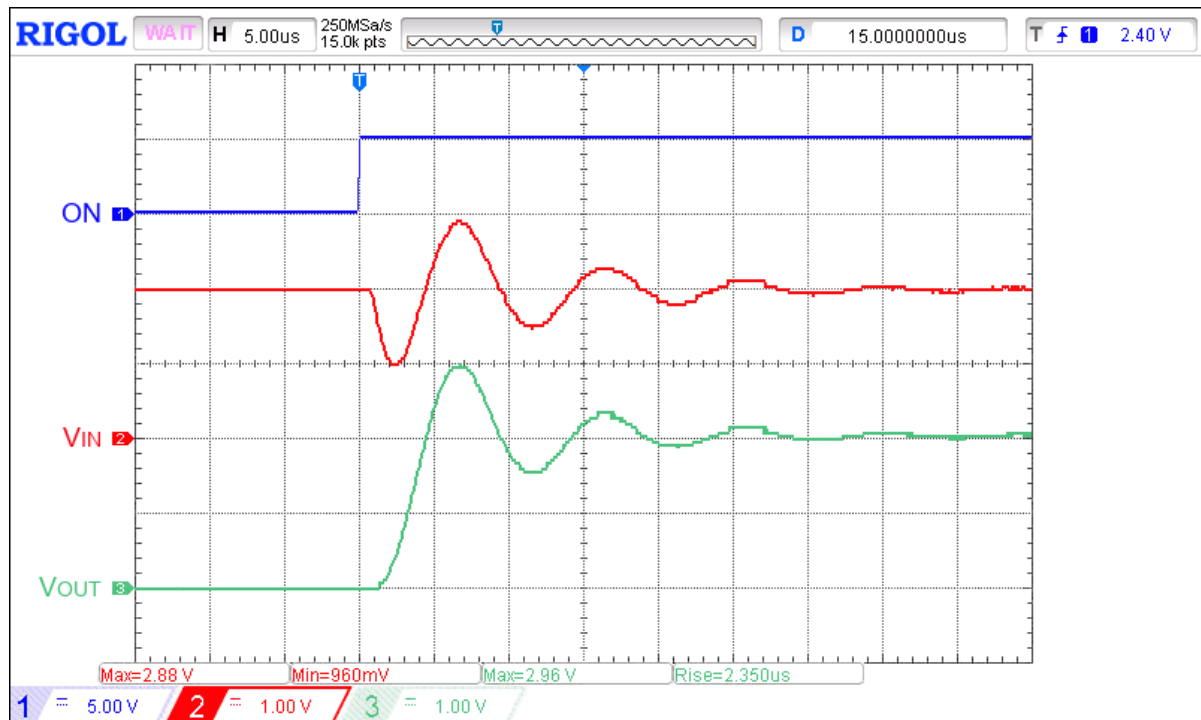


Figure 8. An IPS Powering Up a 1  $\mu$ F Load Capacitor when  $V_{IN}$  is Applied Before  $V_{DD}$ .



## Selecting Input and Output Capacitors with Dialog's Integrated Power Switches

In fast turn-on applications into a capacitive load, some unwanted effects may be observed. One such effect is illustrated in Figure 9, where a Dialog Nanopower IPS powers up a 10  $\mu\text{F}$  load capacitor.



**Figure 9. A Fast, Nanopower IPS Powering Up a 10  $\mu\text{F}$  Load Capacitor.**

This behavior is related to a big inrush current [2], caused by applying a voltage through IPS across a discharged (or an uncharged) capacitor. The resulting inrush current can be calculated by equation below:

$$\text{Inrush Current, } I = C \frac{dV}{dt},$$

where

C - is the total load capacitance;

$\frac{dV}{dt}$  - the IPS's  $V_{\text{OUT}}$  slew rate during voltage ramp up.

This inrush current leads to the voltage drop at  $V_{\text{IN}}$  during IPS power up. Also, this current builds a magnetic field in the parasitic inductance caused by wires from the power supply. When the voltage drop occurs, the magnetic field changes in strength and collapses. This leads to voltage spikes that appear at  $V_{\text{IN}}$  and, respectively, at  $V_{\text{OUT}}$ . These voltage spikes can be much larger than the initial  $V_{\text{IN}}$  voltage level and can greatly shorten the IPS's long-term reliability [2] or even damage it and any other circuit downstream of it.

One way to minimize this effect is to reduce inrush current for a given load capacitance by decreasing (or slowing down) the IPS's  $V_{\text{OUT}}$  slew rate. This can be achieved by using power switches with controlled slew rate [3], [4]. However, this method also leads to increasing total circuit turn-on time.

Another way is to add an input capacitor at  $V_{\text{IN}}$  to minimize the voltage drop during fast power-up events. Figure 10 shows a turn-on operation for a 10  $\mu\text{F}$  load capacitor with a 1000  $\mu\text{F}$  aluminum electrolytic capacitor at  $V_{\text{IN}}$ . As shown, the voltage drop is much smaller, but it is still present due to

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ESR and ESL parasitic elements in capacitors described earlier. In the case of using a capacitor with smaller ESR and ESL, it is possible to get the same or even better results with smaller capacitance value. For example, in Figure 11, a low ESR/ESL aluminum polymer capacitor was used to power up a 10 $\mu$ F load capacitor.

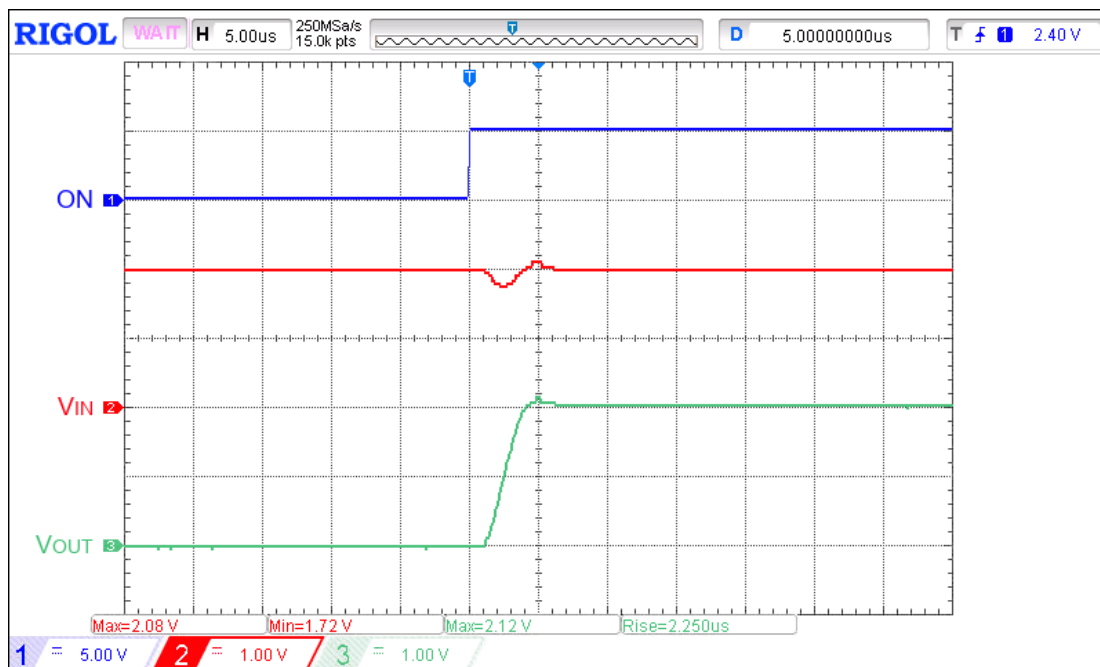


Figure 10. A Fast, Nanopower IPS Powering Up a 10  $\mu$ F Load Capacitor with a 1000  $\mu$ F Aluminum Electrolytic Capacitor at VIN ( $C_{IN}$ ).

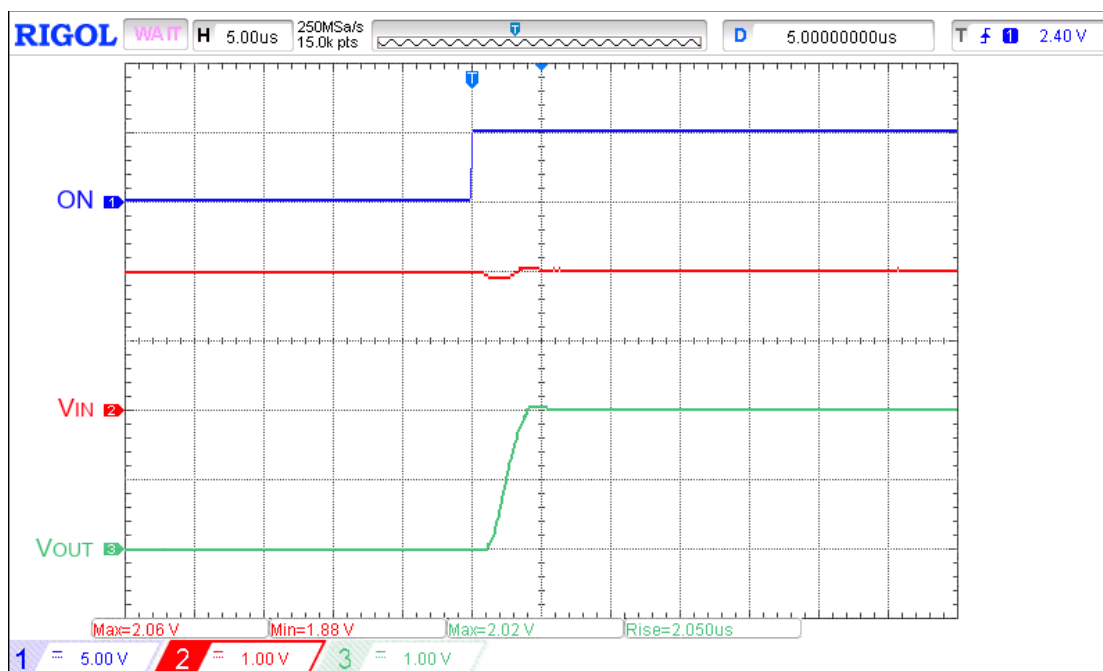
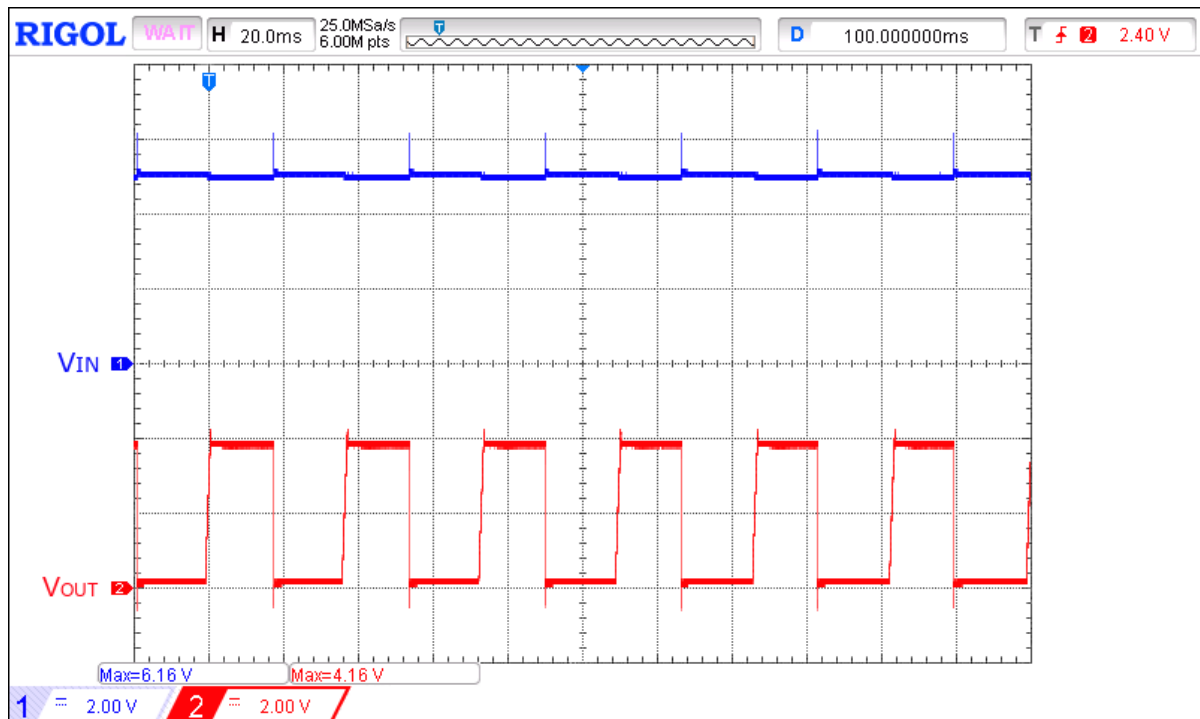


Figure 11. A Fast, Nanopower IPS Powering up a 10  $\mu$ F Load Capacitor with a 270  $\mu$ F Low ESR/ESL Aluminum Polymer Input Capacitor at VIN ( $C_{IN}$ ).

## Selecting Input and Output Capacitors with Dialog's Integrated Power Switches

Yet another situation which requires input and output capacitors is when an IPS's active current limit (ACL) or short-circuit protection (SCL) is triggered. During these events, current through the IPS may change in large steps (eg. suddenly shut off) which if long wires (inductance) are present can cause large voltage spikes (Figure 12) that could damage the IPS and even other components powered from the same power rail. To eliminate these voltage spikes, it is necessary to a) use shorter connections from power supply to the device (Figure 13) and/or b) add or increase the corresponding capacitance (Figure 14). Since long wires tend to be more prevalent on the power supply ( $V_{IN}$ ) side, the examples below show how the voltage spikes in this situation are mitigated.



**Figure 12. The Parasitic Inductance of 1.2-m Length AWG13 Wires from Power Supply Causes Voltage Spikes at  $V_{IN}$  during Active Current Limit Operation.**

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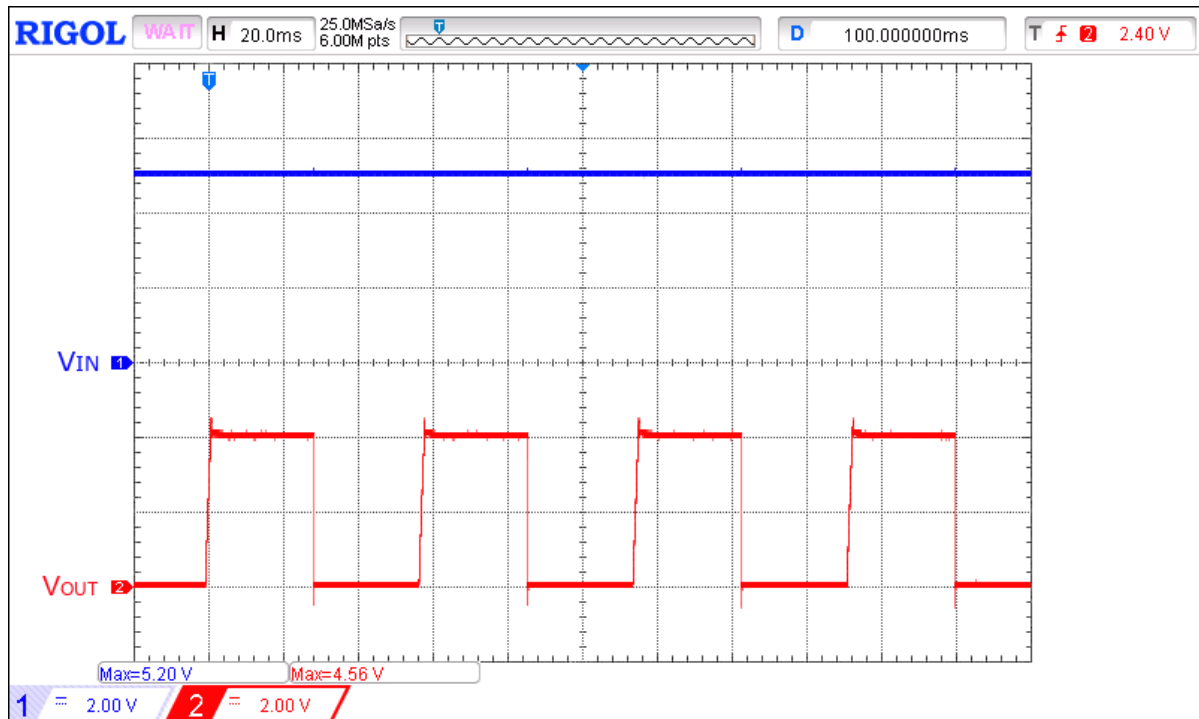


Figure 13. Shorter, 0.2-m Length of AWG13 Wires from Power Supply Do Not Cause Voltage Spikes at VIN during Active Current Limit Operation.

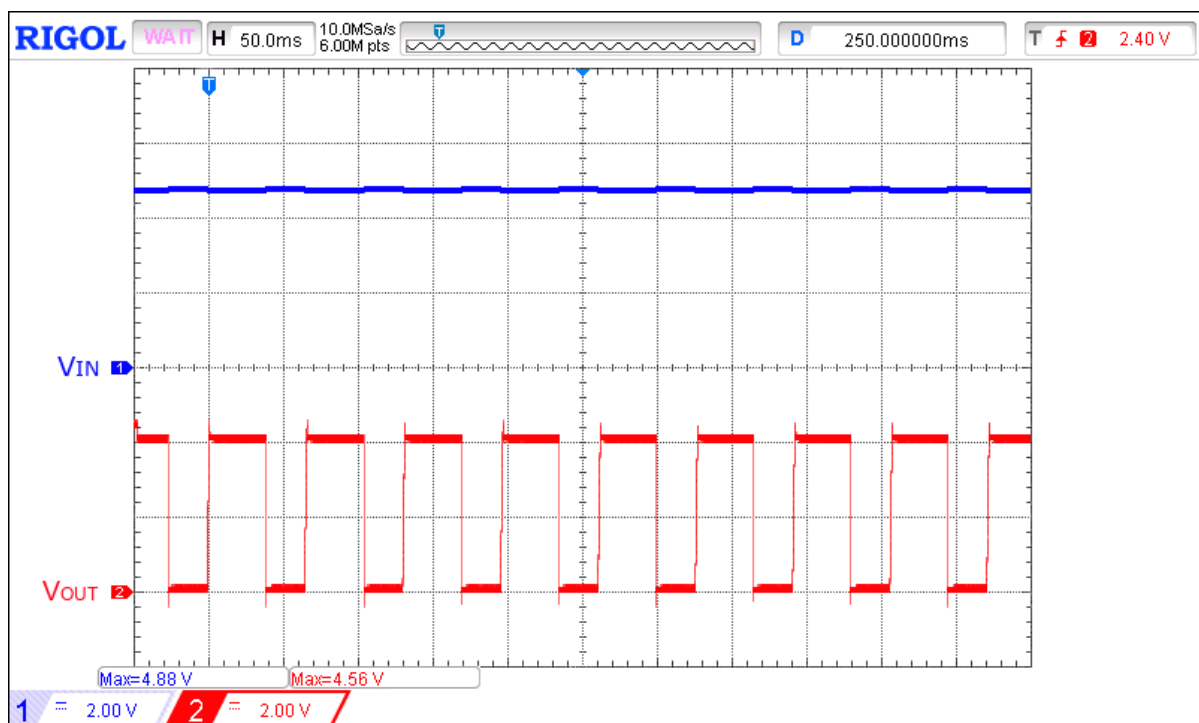


Figure 14. Adding 470  $\mu$ F Low ESR/ESL Aluminum Polymer Capacitor at VIN Eliminates Voltage Spikes Caused by the Parasitic Inductance of the 1.2-m Length AWG13 Wires from Power Supply during Active Current Limit Operation.

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**Selecting Input and Output Capacitors with Dialog's  
Integrated Power Switches****6 Conclusions**

Capacitors are fundamental components in most digital and analog circuits. Dialog IPSs don't inherently require input and output capacitors. However, application requirements may dictate the use of input and output capacitors. In IPS applications where input and output capacitors are needed, ceramic and/or tantalum capacitors are recommended.

## Selecting Input and Output Capacitors with Dialog's Integrated Power Switches

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